

# MICROPROCESSOR CONTROLLED STEREO SOUND GENERATOR FOR SOUND EFFECTS AND MUSIC SYNTHESIS

## GENERAL DESCRIPTION

The SAA1099 is a monolithic integrated circuit designed for generation of stereo sound effects and music synthesis.

### Features

- Six frequency generators
  - eight octaves per generator
  - 256 tones per octave
- Two noise generators
- Six noise/frequency mixers
- Twelve amplitude controllers
- Two envelope controllers
- Two 6-channel mixers/current sink analogue output stages
- TTL input compatible
- Readily interfaces to 8-bit microcontroller
- Minimal peripheral components
- Simple output filtering

### Applications

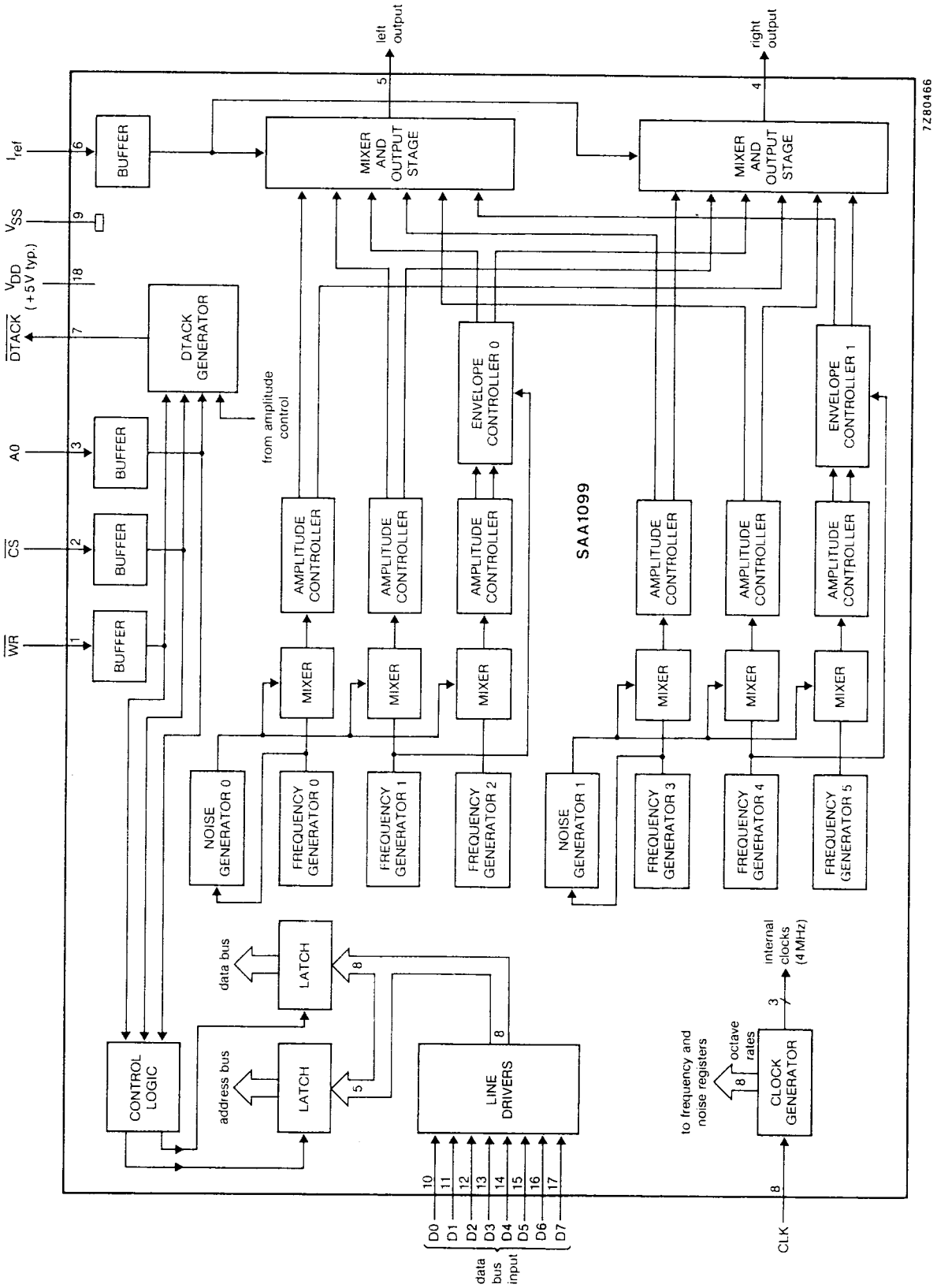
- Consumer games systems
- Home computers
- Electronic organs
- Arcade games
- Toys
- Chimes/alarm clocks

### QUICK REFERENCE DATA

Supply voltage (pin 18)	$V_{DD}$	typ.	5 V
Supply current (pin 18)	$I_{DD}$	typ.	70 mA
Reference current (pin 6)	$I_{ref}$	typ.	250 $\mu$ A
Total power dissipation	$P_{tot}$		500 mW
Operating ambient temperature range	$T_{amb}$		0 to + 70 °C

### PACKAGE OUTLINE

18-lead DIL; plastic (SOT102).



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Fig. 1 Block diagram.

PINNING

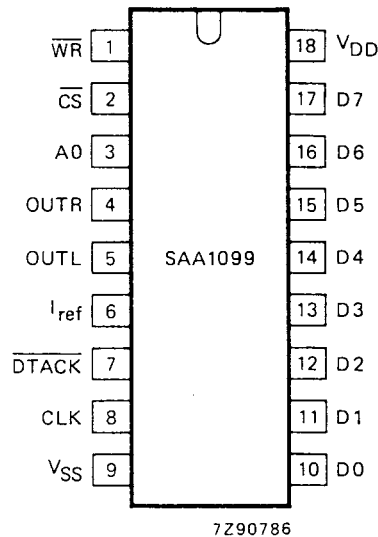


Fig. 2 Pinning diagram.

PIN DESIGNATION

1	$\overline{WR}$	<b>Write Enable:</b> active LOW input which operates in conjunction with $\overline{CS}$ and A0 to allow writing to the internal registers.
2	$\overline{CS}$	<b>Chip Select:</b> active LOW input to identify valid $\overline{WR}$ inputs to the chip. This input also operates in conjunction with $\overline{WR}$ and A0 to allow writing to the internal registers.
3	A0	<b>Control/Address select:</b> input used in conjunction with $\overline{WR}$ and $\overline{CS}$ to load data to the control register (A0 = 0) or the address buffer (A0 = 1).
4	OUTR	<b>Right channel output:</b> a 7-level current sink analogue output for the 'right' component. This pin requires an external load resistor.
5	OUTL	<b>Left channel output:</b> a 7-level current sink analogue output for the 'left' component. This pin requires an external load resistor.
6	$I_{ref}$	<b>Reference current supply:</b> used to bias the current sink outputs.
7	$\overline{DTACK}$	<b>Data Transfer Acknowledge:</b> open drain output, active LOW to acknowledge successful data transfer. On completion of the cycle $\overline{DTACK}$ is set to inactive.
8	CLK	<b>Clock:</b> input for an externally generated clock at a nominal frequency of 8 MHz.
9	VSS	<b>Ground:</b> 0 V.
10-17	D0-D7	<b>Data:</b> Data bus input.
18	VDD	<b>Power supply:</b> + 5 V typical.

## FUNCTIONAL DESCRIPTION

The following sections provide a detailed functional description of the SAA1099 as shown in the block diagram, Fig. 1.

### Frequency generators

Six frequency generators can each select one of 8 octaves and one of 256 tones within an octave. A total frequency range of 31 Hz to 7,81 kHz is available. The outputs may also control noise or envelope generators. All frequency generators have an enable bit which switches them on and off, making it possible to preselect a tone and to make it inaudible when required. The frequency generators may be synchronized using the frequency reset bit.

The frequency ranges per octave are:

Octave	Frequency range
0	31 Hz to 61 Hz
1	61 Hz to 122 Hz
2	122 Hz to 244 Hz
3	245 Hz to 488 Hz
4	489 Hz to 977 Hz
5	978 Hz to 1,95 kHz
6	1,96 kHz to 3,91 kHz
7	3,91 kHz to 7,81 kHz

### Noise generators

The two noise generators both have a programmable output. This may be a software controlled noise via one of the frequency controlled generators or one of three pre-defined noises. There is no tone produced by the frequency generator when it is controlling the noise generator. The noise produced is based on double the frequency generator output, i.e. a range of 61 Hz to 15,6 kHz.

In the event of a pre-defined noise being chosen, the output of noise generator 0 can be mixed with frequency generator 0, 1 and 2; and the output of noise generator 1 can be mixed with frequency generator 3, 4 and 5. In order to produce an equal level of noise and tone outputs (when both are mixed) the amplitude of the tone is increased. The three pre-defined noises are based on a clock frequency of 7,8 kHz, 15,6 kHz or 31,25 kHz.

### Noise/frequency mixers

Six noise/frequency mixers each with four selections

- Channel off
- Frequency only
- Noise only
- Noise and frequency

Each mixer channel has one of the frequency generator outputs fed to it, three channels use noise generator 0 and the other three use noise generator 1.

### Amplitude controllers

Each of the six channel outputs from the mixer is split up into a right and left component giving effectively twelve amplitude controllers. An amplitude of 16 possible levels is assigned to each of the twelve signals. With this configuration a stereo effect can be achieved by varying only the amplitude component. The moving of a sound from one channel to the other requires, per tone, only one update of the amplitude register contents.

When an envelope generator is used, the amplitude levels are restricted. The number of levels available is then reduced to eight. This is achieved by disabling the least significant bit (LSB) of the amplitude control.

### Envelope controllers

Two of the six tone generators are under envelope control. This applies to both the left and right outputs from the tone generator.

The envelope has the following eight possible modes:

- Amplitude is zero
- Single attack
- Single decay
- Single attack-decay (triangular)
- Maximum amplitude
- Continuous attack
- Continuous decay
- Continuous attack-decay

The timing of the envelope controllers is programmable using one of the frequency generators (see Fig. 1). When the envelope mode is selected for a channel its control resolution is halved for that channel from 16 levels to 8 levels by rounding down to the nearest even level.

There is also the capability of controlling the 'right' component of the channel with inverse of the 'left' component, which remains as programmed.

A direct enable permits the start of an envelope to be defined, and also allows termination of an envelope at any time. The envelope rate may be controlled by a frequency channel (see Fig. 1), or by the microprocessor writing to the address buffer register. If the frequency channel controlled is OFF ( $NE = FE = 0$ ) the envelope will appear at the output, which provides an alternative 'non-square' tone capability. In this event the frequency will be the envelope rate, which provided the rate is from the frequency channel, will be a maximum of 1 kHz. Higher frequencies of up to 2 kHz can be obtained by the envelope resolution being halved from 16 levels to 8 levels. Rates quoted are based on the input of a 8 MHz clock.

### Six-channel mixers/current sink analogue output stages

Six channels are mixed together by the two mixers allowing each one to control one of six equally weighted current sinks, to provide a seven level analogue output.

### Command/control select

In order to simplify the microprocessor interface the command and control information is multiplexed. To select a register in order to control frequencies, amplitudes, etc. the command-register has to be loaded. The contents of this register determines to which register the data is written in the next control-cycle. If a continuous update of the control-register is necessary, only the control-information has to be written (the command-information does not change).

If the command/control select (A0) is logic 0, the byte transfer is control; if A0 is logic 1, the byte transfer is command.

### Interface to microprocessor

The SAA1099 is a data bus based I/O peripheral. Depending on the value of the command/control signal (A0) the  $\overline{CS}$  and  $\overline{WR}$  signals control the data transfer from the microprocessor to the SAA1099. The data-transfer-acknowledge ( $\overline{DTACK}$ ) indicates that the data transfer is completed. When, during the write cycle, the microprocessor recognizes the  $\overline{DTACK}$ , the bus cycle will be completed by the processor.

## RATINGS

Limiting values in accordance with the Absolute Maximum System (IEC 134)

Supply voltage (pin 18)	$V_{DD}$	-0,3 to + 7,5 V
Maximum input voltage	$V_I$	-0,3 to + 7,5 V
at $V_{DD} = 4,5$ to $5,5$ V	$V_I$	-0,5 to + 7,5 V
Maximum output current	$I_O$	max. 10 mA
Total power dissipation	$P_{tot}$	500 mW
Storage temperature range	$T_{stg}$	-55 to + 125 °C
Operating ambient temperature range	$T_{amb}$	0 to + 70 °C
Electrostatic handling*	$V_{es}$	-1000 to + 1000 V

\* Equivalent to discharging a 250  $\mu$ F capacitor through a 1 k $\Omega$  series resistor.

**D.C. CHARACTERISTICS**

$V_{DD} = 5\text{ V} \pm 10\%$ ;  $T_{amb} = 0\text{ to }70\text{ }^{\circ}\text{C}$ ; unless otherwise specified

parameter	symbol	min.	typ.	max.	unit
<b>Supply</b>					
Supply voltage	$V_{DD}$	4,5	5,0	5,5	V
Supply current	$I_{DD}$	—	70	100	mA
Reference current (note 1)	$I_{ref}$	100	250	400	$\mu\text{A}$
<b>INPUTS</b>					
Input voltage HIGH	$V_{IH}$	2,0	—	6,0	V
Input voltage LOW	$V_{IL}$	-0,5	—	0,8	V
Input leakage current	$\pm I_{LI}$	—	—	10	$\mu\text{A}$
Input capacitance	$C_I$	—	—	10	pF
<b>OUTPUTS</b>					
$\overline{DTACK}$ (open drain; note 2)					
Output voltage LOW at $I_{OL} = 3,2\text{ mA}$	$V_{OL}$	0	—	0,4	V
Voltage on pin 7 (OFF state)	$V_{7-9}$	-0,3	—	6,0	V
Output capacitance (OFF state)	$C_O$	—	—	10	pF
Load capacitance	$C_L$	—	—	150	pF
Output leakage current (OFF state)	$-I_{LO}$	—	—	10	$\mu\text{A}$
<b>Audio outputs (pins 4 and 5)</b>					
<i>With fixed <math>I_{ref}</math> (note 3)</i>					
One channel on	$I_{O1}/I_{ref}$	90	—	120	%
Six channels on	$I_{O6}/6 \times I_{ref}$	85	—	110	%
<i>With <math>I_{ref} = 250\text{ } \mu\text{A}</math>; <math>R_L = 1,5\text{ k}\Omega</math> (<math>\pm 5\%</math>)</i>					
One channel on	$I_{O1}/I_{ref}$	90	—	110	%
Six channels on	$I_{O6}/6 \times I_{ref}$	85	—	105	%
Output current one channel on	$I_{O1}$	225	—	275	$\mu\text{A}$
Output current six channels on	$I_{O6}$	1,3	—	1,6	mA
<i>With resistor supplying <math>I_{ref}</math> (note 4)</i>					
Output current one channel on	$I_{O1}$	150	—	350	$\mu\text{A}$
Output current six channels on	$I_{O6}$	0,9	—	1,9	mA
Load resistance	$R_L$	600	—	—	$\Omega$
D.C. leakage current all channels off	$-I_{LO}$	—	—	10	$\mu\text{A}$
Maximum current difference between left and right current sinks (note 5)	$\pm I_{Omax}$	—	—	15	%
Signal-to-noise ratio (note 6)	S/N	—	tbf	—	dB

Vf

## A.C. CHARACTERISTICS

$V_{DD} = 5\text{ V} \pm 10\%$ ;  $T_{amb} = 0\text{ to }70\text{ }^{\circ}\text{C}$ ; timing measurements taken at 2,0 V for a logic 1 and 0,8 V for a logic 0 unless otherwise specified (see waveforms Figs 3 and 4)

parameter	symbol	min.	typ.	max.	unit
<b>Bus interface timing (see Fig. 3)</b>					
A0 set-up time to $\overline{\text{CS}}$ fall	$t_{ASC}$	0	—	—	ns
$\overline{\text{CS}}$ LOW to $\overline{\text{WR}}$ fall	$t_{CSW}$	30	—	—	ns
A0 set-up time to $\overline{\text{WR}}$ fall	$t_{ASW}$	50	—	—	ns
$\overline{\text{WR}}$ LOW time	$t_{WL}$	100	—	—	ns
Data bus valid to $\overline{\text{WR}}$ rise	$t_{BSW}$	100	—	—	ns
$\overline{\text{DTACK}}$ fall delay from $\overline{\text{WR}}$ fall (note 7)	$t_{DFW}$	0	—	85	ns
A0 hold time from $\overline{\text{WR}}$ HIGH	$t_{AHW}$	0	—	—	ns
$\overline{\text{CS}}$ hold time from $\overline{\text{WR}}$ HIGH	$t_{CHW}$	0	—	—	ns
Data bus hold time from $\overline{\text{WR}}$ HIGH	$t_{DHW}$	0	—	—	ns
$\overline{\text{DTACK}}$ rise delay from $\overline{\text{WR}}$ HIGH	$t_{DRW}$	0	—	100	ns
Bus cycle time (note 8)	$t_{CY}$	$4t_{CLK}$	—	—	
Bus cycle time (note 9)	$t_{CY}$	$16t_{CLK}$	—	—	
<b>Clock input timing (see Fig. 4)</b>					
Clock period	$t_{CLK}$	120	125	255	ns
Clock LOW time	$t_{LOW}$	55	—	—	ns
Clock HIGH time	$t_{HIGH}$	55	—	—	ns

## Notes to the characteristics

- Using an external constant current generator to provide a nominal  $I_{ref}$  or external resistor connected to  $V_{DD}$ .
- This output is short-circuit protected to  $V_{DD}$  and  $V_{SS}$ .
- Measured with  $I_{ref}$  a constant value between 100 and 400  $\mu\text{A}$ ; load resistance ( $R_L$ ) allowed to match E12 (5%) in all applications via:

$$R_L = 0,6 [I_{ref}]^{-1} - 16 [I_{ref}]^{-0,5} \pm 12\%$$

- Measured with  $R_{ref} = 10\text{ k}\Omega$  ( $\pm 5\%$ ) connected between  $I_{ref}$  and  $V_{DD}$ ;  $R_L = 1,5\text{ k}\Omega$  ( $\pm 5\%$ ); OUTR and OUTL short-circuit protected to  $V_{SS}$ .
- Left and right outputs must be driven with identical configuration.
- Sample tested value only.
- This timing parameter only applies when no wait states are required; otherwise parameter is invalid.
- The minimum bus cycle time of four clock periods is for loading all registers except the amplitude registers.
- The minimum bus cycle time of 16 clock periods is for loading the amplitude registers. In a system using  $\overline{\text{DTACK}}$  it is possible to achieve minimum times of 500 ns. Without  $\overline{\text{DTACK}}$  the parameter given must be used.



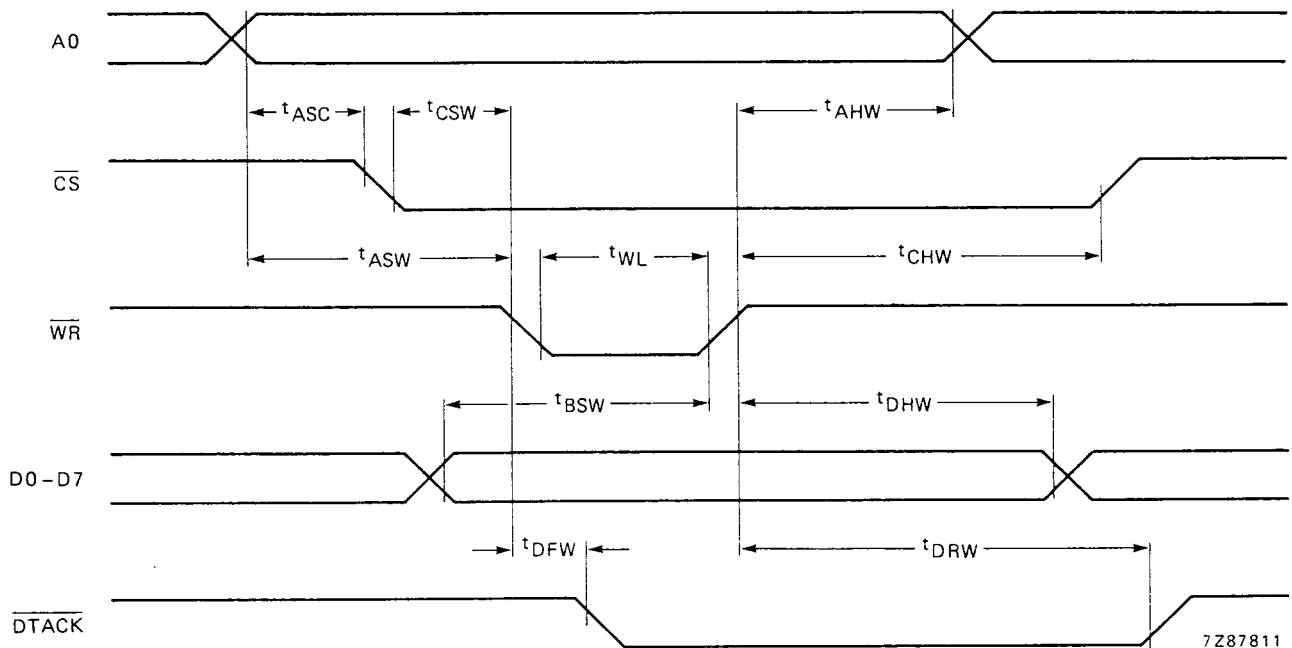


Fig. 3 Bus interface waveforms.

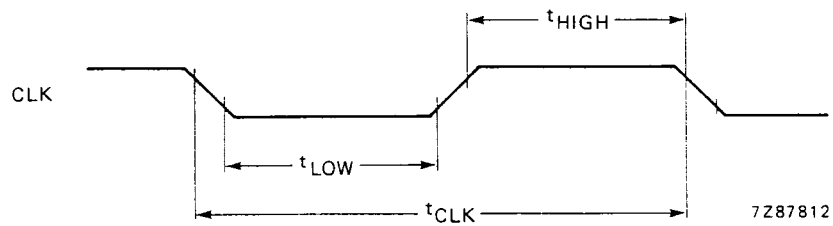


Fig. 4 Clock input waveform.

## APPLICATION INFORMATION

### Device operation

The SAA1099 uses pulse width modulation to achieve amplitude and envelope levels. The twelve signals are mixed in an analogue format (6 'left' and 6 'right') before leaving the chip. The amplitude and envelope signals chop the output at a minimum rate of 62,5 kHz, compared with the highest tone output of 7,81 kHz. Simple external low-pass filtering is used to remove the high frequency components.

Rates quoted are based on the input of a 8 MHz clock.

A data bus based write only structure is used to load the on-board registers. The data bus is used to load the address for a register, and subsequently the data to that register. Once the address is loaded multiple data loads to that register can be performed.

The selection of address or data is made by the single address bit A0, as shown in register maps Table 1 and Table 2.

The bus control signals  $\overline{WR}$  and  $\overline{CS}$  are designed to be compatible with a wide range of microprocessors, a  $\overline{DTACK}$  output is included to optimise the interface with an S68000 series microprocessor. In most bus cycles  $\overline{DTACK}$  will be returned immediately, this applies to all register address load cycles and all except amplitude data load cycles. With respect to amplitude data, a number of wait cycles may need to be performed, depending on the time since the previous amplitude load.  $\overline{DTACK}$  will indicate the number of required waits.

### Register description (see Tables 2 and 3)

The amplitudes are assigned with 'left' and 'right' components in the same byte, on a channel by channel basis. The spare locations that are left between blocks of registers is to allow for future expansion, and should be written as zero's. The tone within an octave is defined by eight bits and the octave by three bits. Note that octaves are paired (0/1, 2/3 etc.). The frequency and noise enables are grouped together for ease of programming. The controls for noise 'colour' (clock rate) are grouped in one byte.

The envelope registers are positioned in adjacent locations. There are two types of envelope controls, direct acting controls and buffered controls. The direct acting controls always take immediate effect, and are:

- Envelope enable (reset)
- Envelope resolution (16/8 level)

The buffered controls are acted upon only at the times shown in Fig. 5 and control selection of:

- Envelope clock source
- Waveform type
- Inverted/non-inverted 'right' component

**Table 1 External memory map**

select A0	data bus inputs								operations
	D7	D6	D5	D4	D3	D2	D1	D0	
0	D7	D6	D5	D4	D3	D2	D1	D0,	data for internal registers internal register address
1	X	X	X	A4	A3	A2	A1	A0	

Where X = don't care state.

Table 2 Internal register map

register address	data bus inputs								operations
	D7	D6	D5	D4	D3	D2	D1	D0	
00	AR03	AR02	AR01	AR00	AL03	AL02	AL01	AL00	amplitude 0 right channel; left channel
01	1	1	1	1	1	1	1	1	amplitude 1 right/left
02	2	2	2	2	2	2	2	2	amplitude 2 right/left
03	3	3	3	3	3	3	3	3	amplitude 3 right/left
04	4	4	4	4	4	4	4	4	amplitude 4 right/left
05	5	5	5	5	5	5	5	5	amplitude 5 right/left
06	X	X	X	X	X	X	X	X	
07	X	X	X	X	X	X	X	X	
08	F07	F06	F05	F04	F03	F02	F01	F00	frequency of tone 0
09	1	1	1	1	1	1	1	1	frequency of tone 1
0A	2	2	2	2	2	2	2	2	frequency of tone 2
0B	3	3	3	3	3	3	3	3	frequency of tone 3
0C	4	4	4	4	4	4	4	4	frequency of tone 4
0D	F57	F56	F55	F54	F53	F52	F51	F50	frequency of tone 5
0E	X	X	X	X	X	X	X	X	
0F	X	X	X	X	X	X	X	X	
10	X	012	011	010	X	002	001	000	octave 1; octave 0
11	X	032	031	030	X	022	021	020	octave 3; octave 2
12	X	052	051	050	X	042	041	040	octave 5; octave 4
13	X	X	X	X	X	X	X	X	
14	X	X	FE5	FE4	FE3	FE2	FE1	FE0	frequency enable
15	X	X	NE5	NE4	NE3	NE2	NE1	NE0	noise enable
16	X	X	N11	N10	X	X	N01	N00	noise generator 1; noise generator 0
17	X	X	X	X	X	X	X	X	
18	E07	X	E05	E04	E03	E02	E01	E00	envelope generator 0
19	E17	X	E15	E14	E13	E12	E11	E10	envelope generator 1
1A	X	X	X	X	X	X	X	X	
1B	X	X	X	X	X	X	X	X	
1C	X	X	X	X	X	X	RST	SE	frequency reset (all channels) sound enable (all channels)
1D	X	X	X	X	X	X	X	X	
1E	X	X	X	X	X	X	X	X	
1F	X	X	X	X	X	X	X	X	

Where:

All don't cares (X) should be written as zero's.

00 to 1F block of registers repeats eight times in the block between addresses 00 to FF (full internal memory map).

## APPLICATION INFORMATION (continued)

Table 3 Register description

bit	description
ARn3; ARn2; ARn1; ARn0 (n = 0,5)	4 bits for amplitude control of right channel 0 0 0 0 minimum amplitude (off) 1 1 1 1 maximum amplitude
ALn3; ALn2; ALn1; ALn0 (n = 0,5)	4 bits for amplitude control of left channel 0 0 0 0 minimum amplitude (off) 1 1 1 1 maximum amplitude
Fn7 to Fn0 (n = 0,5)	8 bits for frequency control of the six frequency generators 0 0 0 0 0 0 0 0 lowest frequency 1 1 1 1 1 1 1 1 highest frequency
On2; On1; On0 (n = 0,5)	3 bits for octave control 0 0 0 lowest octave (31 Hz to 61 Hz) 0 0 1 (61 Hz to 122 Hz) 0 1 0 (122 Hz to 244 Hz) 0 1 1 (245 Hz to 488 Hz) 1 0 0 (489 Hz to 977 Hz) 1 0 1 (978 Hz to 1,95 kHz) 1 1 0 (1,96 kHz to 3,91 kHz) 1 1 1 highest octave (3,91 kHz to 7,81 kHz)
FEn (n = 0,5)	frequency enable bit (one tone per generator) FEn = 0 indicates that frequency 'n' is off
NEn (n = 0,5)	noise enable bit (one tone per generator) NEn = 0 indicates that noise 'n' is off
Nn1; Nn0 (n = 0,1)	2 bits for noise generator control. These bits select the noise generator rate (noise 'colour') Nn1 Nn0 clock frequency 0 0 31,3 kHz 0 1 15,6 kHz 1 0 7,6 kHz 1 1 61 Hz to 15,6 kHz (frequency generator 0/3)

bit	description
En7; En5 to En0 (n = 0,1)	<p>7 bits for envelope control</p> <p>En0 0 left and right component have the same envelope 1 right component has inverse of envelope that is applied to left component</p> <p>En3 En2 En1</p> <p>0 0 0 zero amplitude 0 0 1 maximum amplitude 0 1 0 single decay 0 1 1 repetitive decay 1 0 0 single triangular 1 0 1 repetitive triangular 1 1 0 single attack 1 1 1 repetitive attack</p> <p>En4 0 4 bits for envelope control (maximum frequency = 977 Hz) 1 3 bits for envelope control (maximum frequency = 1,95 kHz)</p> <p>En5 0 internal envelope clock (frequency generator 1 or 4) 1 external envelope clock (address write pulse)</p> <p>En7 0 reset (no envelope control) 1 envelope control enabled</p>
SE	<p>SE sound enable for all channels (reset on power-up to 0)</p> <p>0 all channels disabled 1 all channels enabled</p>
RST	<p>Reset signal to all frequency generators</p> <p>0 all generators enabled 1 all generators reset and synchronized</p>

**Note**

All rates given are based on the input of a 8 MHz clock.

APPLICATION INFORMATION (continued)

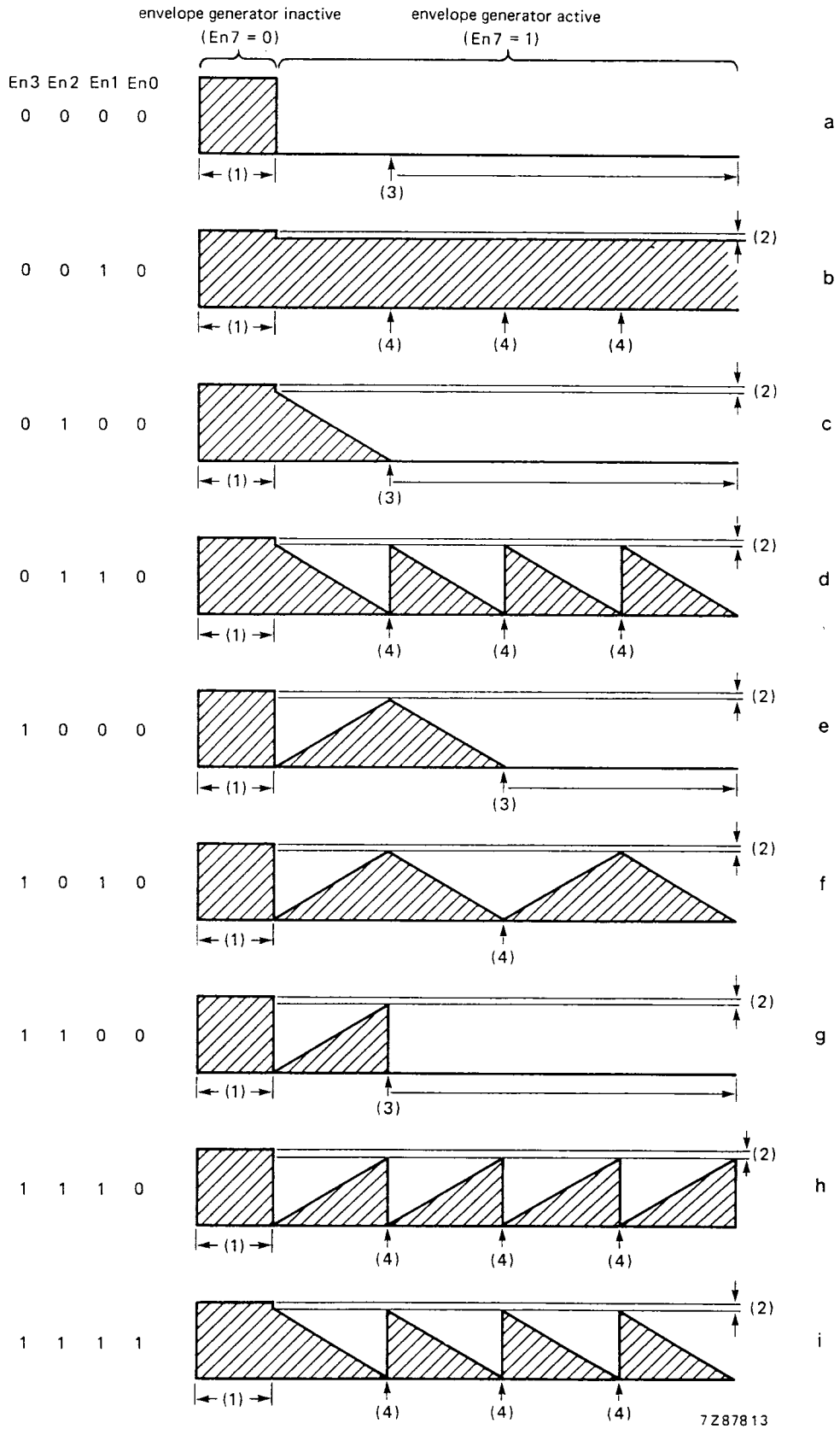


Fig. 5 Envelope waveforms.

Notes to Fig. 5

- (1) The level at this time is under amplitude control only ( $En7 = 0$ ; no envelope).
- (2) When the generator is active ( $En7 = 1$ ) the maximum level possible is 7/8ths of the amplitude level.
- (3) After position (3) the buffered controls will be acted upon when loaded.
- (4) At positions (4) the buffered controls will be acted upon if already loaded.
- (5) Waveforms 'a' to 'h' show the left channel ( $En0 = 0$ ; left and right components have the same envelope).  
Waveform 'i' shows the right channel ( $En0 = 1$ ; right component inverse of envelope applied to left).

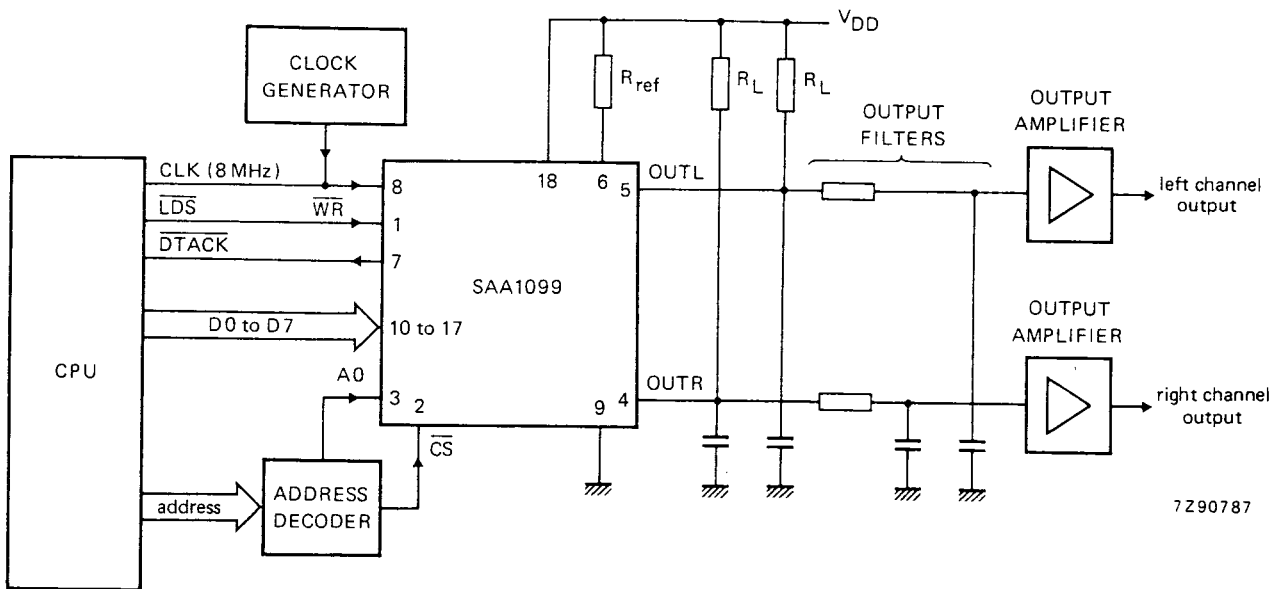


Fig. 6 Typical application circuit diagram.